# **CSE311 Microwave Engineering**



# LEC (03) – Chapter (03) Lec (03) General Plane Wave Solution



Assoc. Prof. Dr. Moataz Elsherbini

motaz.ali@feng.bu.edu.eg

# **Chapter 2 Contents**

#### Review on:

**Maxwell's Equations for Time Harmonic Fields** 

- 2.1 Introduction
- 2.2 General Plane Wave Equations
- 2.3 Solution of Wave Equations in a General Medium
- 2.4 Solution of Wave Equations in Free Space
- 2.5 Solution of Wave Equations in a Lossless Medium
- 2.6 Solution of Wave Equations in a Good Conductor

# 1.4 Maxwell's Equations

#### 1.4.1 Maxwell's Equations in Point or Differential Form

We have already obtained two of the Maxwell's equations for time-varying fields which are:

$$\nabla x \vec{E} = -\frac{\partial \vec{B}}{\partial t}$$
 Faraday's Law of induction (1.15)  
 
$$\nabla x \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t}$$
 Ampere's Circuital Law (1.16)

The remaining two equations are unchanged from their non-time-varying from which are:

$$abla . \vec{D} = 
ho_v$$
 Gauss's Law for Electric Field (1.17)  $abla . \vec{B} = 0$  Gauss's Law for Magnetic Field (1.18)

$$\nabla . \vec{B} = 0$$
 Gauss's Law for Magnetic Field (1.18)

Equation (1.18) acknowledges the fact that "magnetic charges" or poles are not known to exist. Magnetic flux is always found in closed loops and never diverges from a point source.

$$\vec{D} = \varepsilon \vec{E}$$
  $\vec{B} = \mu \vec{H}$   $\vec{J} = \sigma \vec{E}$ 

#### 2.1 Introduction

- Similarly,  $\varepsilon_0$ = 8.854 x10<sup>-12</sup> = (1/36 $\pi$ )x10<sup>-9</sup> F/m) is the permittivity of free-space and  $\varepsilon_r$  > 1 is the relative permittivity of the material.
- Loss tangent values will determine types of media, the loss tangent is defined as:  $\tan \delta = \sigma / \omega \varepsilon$
- Loss tangent values will determine types of media:
  - $\tan \delta$  small ( $\sigma / \omega \epsilon < 0.1$ ) good dielectric
  - $\tan \delta \log (\sigma / \omega \epsilon > 10)$   $\frac{10}{3}$
- Another factor that determined the characteristic of the media is operating frequency, f or the angular frequency  $\omega = 2\pi f$ . A medium can be regarded as a good conductor at low frequency might be a good dielectric at higher frequency.
- In this chapter, our major goals are to:
- 1. Use Maxwell's equations to derive the wave equations in case of the medium is free of charge,  $\rho_v$  = 0 and then solve the wave equations to determine the wave characteristics and parameters in the following media:
  - a. General medium ( $\sigma \neq 0$ ,  $\mu = \mu_r \mu_o$  and  $\varepsilon = \varepsilon_r \varepsilon_o$  and  $\sigma \approx \omega \varepsilon$ ) lossy medium.
  - b. Free space or air ( $\sigma = 0$ ,  $\mu = \mu_o$  and  $\varepsilon = \varepsilon_o$ ).
  - b. Lossless (perfect) dielectric or medium ( $\sigma = 0$ ,  $\mu = \mu_r \mu_o$  and  $\varepsilon = \varepsilon_r \varepsilon_o$ ).
  - c. Lossy (imperfect) dielectric or medium ( $\sigma \neq 0$ ,  $\mu = \mu_r \mu_o$ ,  $\epsilon = \epsilon_r \epsilon_o$  and  $\sigma << \omega \epsilon$ ).
  - d. Good conductor medium ( $\sigma \neq 0$ ,  $\mu = \mu_r \mu_o$ ,  $\epsilon = \epsilon_r \epsilon_o$  and  $\sigma >> \omega \epsilon$ ).

# 2.2 General Plane Wave Equations

- The analytical treatment of the uniform plane wave is consider when the wave propagates in dielectric material of conductivity,  $\sigma$ , permittivity,  $\varepsilon = \varepsilon_r \varepsilon_0$ and permeability,  $\mu = \mu_r \mu_o$ .
- The medium is assumed to be homogenous (having constant μ and ε with position) and isotropic (in which  $\mu$  and  $\epsilon$  are invariant with the field orientation). Assume the medium is free of charge, so  $\rho_v = 0$ , under these conditions, Maxwell's equations in point form may be written in terms of E and  $\vec{H}$  only as:

$$\nabla x \vec{E} = -\frac{\partial \vec{B}}{\partial t} = -\mu \frac{\partial \vec{H}}{\partial t} \tag{2.1}$$

$$\nabla x \vec{H} = \vec{J} + \frac{\partial D}{\partial t} = \sigma \vec{E} + \varepsilon \frac{\partial E}{\partial t}$$
 (2.2)

To H of Hy as:  

$$\nabla x \vec{E} = -\frac{\partial \vec{B}}{\partial t} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\nabla x \vec{H} = \vec{J} + \frac{\partial \vec{D}}{\partial t} = \sigma \vec{E} + \varepsilon \frac{\partial \vec{E}}{\partial t}$$

$$\nabla . \vec{E} = \frac{\rho_{\nu}}{\varepsilon} = 0$$
(2.1)
$$\nabla . \vec{H} = 0$$
(2.2)

 $\vec{E}$  is the electric field intensity, in volts per meter (V/m).

His the magnetic field intensity, in amperes per meter (A/m).

is the electric flux density, in coulombs per meter squared (C/m<sup>2</sup>).

is the magnetic flux density, in webers per meter squared (Wb/m²).

is the electric current density, in amperes per meter squared (A/m<sup>2</sup>).

 $\rho_{v}$  is the electric volume charge density, in coulombs per meter cubed (C/m<sup>3</sup>)

$$\frac{\partial \vec{E}}{\partial t} = j\omega \vec{E} \qquad \qquad \frac{\partial \vec{H}}{\partial t} = j\omega \vec{H}$$

# 2.2 General Plane Wave Equations (Continued)

In charge free, linear isotropic, homogenous region, applying Eq.(2.9), the Maxwell's equations [(2.1) to (2.4)] in the Phasor form are:

$$\nabla x \vec{E}_{s} = -j\omega \mu \vec{H}_{s} \tag{2.10}$$

$$\nabla x \vec{H}_{s} = \sigma \vec{E}_{s} + j\omega \varepsilon \vec{E}_{s} = (\sigma + j\omega \varepsilon) \vec{E}_{s}$$
(2.11)

$$\nabla \cdot \vec{E}_{s} = \frac{\rho_{v}}{1} = 0 \tag{2.12}$$

$$\nabla \cdot \vec{E}_{s} = \frac{\rho_{v}}{\mathcal{E}} = 0$$

$$\nabla \cdot \vec{H}_{s} = 0$$
(2.12)
$$(2.13)$$

- Equations from (2.10) to (2.13) constitute the two unknowns,  $\vec{E}$  and  $\vec{H}$ . It required to solve only for one variable, either  $\vec{E}$  or  $\vec{H}$  as follows:
- 1. Take the Curl operator of both sides of (2.10), we have:

$$\nabla x \nabla x \vec{E}_s = -j\omega \mu_o \nabla x \vec{H}_s \tag{2.14}$$

2. Apply the curl identity described in Eq.(A.5) we have:  $\nabla^2 \vec{E} = \nabla(\nabla \cdot \vec{E}) - \nabla \chi \nabla \chi \vec{E}$ 

$$\nabla x \nabla x \vec{E}_s = \nabla (\nabla \cdot \vec{E}_s) - \nabla^2 \vec{E}_s \tag{2.15a}$$

From (2.12), we get:

$$\nabla x \nabla x \vec{E}_s = 0 - \nabla^2 \vec{E}_s = -\nabla^2 \vec{E}_s \tag{2.15b}$$

# 2.2 General Plane Wave Equations (Continued)

• From (2.11) and (2.15) substitute (2.14), the resulting wave equation for  $\vec{E}$  then becomes:

$$\nabla^{2}\vec{E}_{s} + \omega^{2}\mu\varepsilon(1 - j\frac{\sigma}{\omega\varepsilon})\vec{E}_{s} = 0$$
(2.16a)

$$\nabla^2 \vec{E}_s - \gamma^2 \vec{E}_s = 0$$
 (2.16b)

Where γ is the wave number or complex propagation constant of the medium. In this case γ is a function of the material properties, as described by μ, ε and σ. γ is a complex value as it is referred to as complex propagation constant for the medium defined by:

$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{\sigma}{\omega\varepsilon}}$$
(2.16c)

- Where α [Np/m] is called the attenuation constant or coefficient and β [rad/m] is the phase constant or coefficient.
- Eq. (2.16) is the wave equation, or Helmholtz equation, for  $\vec{E}$ .
- An identical wave equation for  $\vec{H}$  can be derived in the same manner as:

$$\nabla^{2} \vec{H}_{s} + \omega^{2} \mu \varepsilon (1 - j - \sigma) \vec{H}_{s} = 0$$
 (2.17a)

$$\nabla^2 \vec{H}_s - \gamma^2 \vec{H}_s = 0 \tag{2.17b}$$

 Equations (2.16) and (2.17) are called the general plane wave equations or Helmholtz equations whose solutions describes the wave behavior and characteristics in a specific media as will be explained in the next sections.

# 2.3 Solution of Wave Equations in a General Medium

#### 2.3.1 Classification of wave solution

• It is now possible to separate the solutions of the wave equation in a charge free region into three basic types of fields. These are:

#### 1. Transverse Electro-Magnetic wave (TEM wave):

It is characterized by the condition that both the electric and magnetic field vectors lie in a plane perpendicular to the direction of propagation, i.e. having no components in the direction of propagation.

#### 2. Transverse Electric wave (TE, or H-wave):

It is characterized by having an electric field which is entirely in a plane transverse to the (assumed) direction of propagation. Only the magnetic field  $\vec{H}$  has a component in the direction of propagation and hence this wave type is also known as H-waves. For TE waves, it is possible to express all field components in terms of the axial magnetic field component.

#### 3. Transverse Magnetic wave (TM, or E-wave)

It is characterized by having a magnetic field which in entirely in a plane transverse to the (assumed) direction of propagation. Only the electric field  $\vec{E}$  has a component in the direction of propagation, and hence this wave type is also known as E-wave. For TM-wave, it is possible to express all field components in term of the axial electric field component.

# 2.3 Solution of Wave Equations in a General Medium (continued) 2.3.2 Solution of Wave Equation

A basic plane wave solution to the above equations can be found by considering an electric field with only an â<sub>x</sub> component and uniform (no variation) in the x and y, then we get:

The Laplacian operator,  $\nabla^2$  is defined in Eq. (A.6) as:  $\nabla^2 V = \frac{\partial^2 V}{\partial x^2} + \frac{\partial^2 V}{\partial y^2} + \frac{\partial^2 V}{\partial z^2}$ 

- The Helmholtz or wave equation for  $\vec{E}$  (2.16) is reduced to:

$$\frac{\partial^2 \vec{E}_{xs}(z)}{\partial z^2} - \gamma^2 \vec{E}_{xs}(z) = 0 \tag{2.19}$$

 The solution to Eq.(2.19) will be forward and backward propagating waves having the phasor solutions in the form:

$$E_{rs}(z) = E^{+}e^{-\gamma z} + E^{-}e^{\gamma z} = E^{+}e^{-\alpha z}e^{-j\beta z} + E^{-}e^{\alpha z}e^{j\beta z}$$
(2.20)

- Where E<sup>+</sup> and E<sup>-</sup> are arbitrary constants.
- The positive traveling wave in phasor form after substituting for  $\gamma$  is:

$$E_{rs}(z) = E^{+}e^{-\gamma z} = E^{+}e^{-\alpha z}e^{-j\beta z}$$
 (2.21)

Multiplying (2.21) by e<sup>jo</sup> and taking the real part yields a form of the field in time domain that can be visualized as:

$$E_{x}(z,t) = E^{+}e^{-\alpha z}\cos(\omega t - \beta z)$$
 (2.22)

#### 2.3.2 Solution of Wave Equation

 The above solution is for time harmonic case at frequency ω. In time domain, this result is written as:

$$E_{x}(z,t) = \operatorname{Re}\left[E_{xs}(z)e^{j\omega t}\right] = \operatorname{Re}\left[\left(E^{+}e^{-\alpha z}e^{-j\beta z} + E^{-}e^{\alpha z}e^{j\beta z}\right)e^{j\omega t}\right]$$

$$E_{x}(z,t) = E^{+}e^{-\alpha z}\cos(\omega t - \beta z) + E^{-}e^{\alpha z}\cos(\omega t + \beta z) \tag{2.23}$$

- From Eqs. (2.22 and 23), it is seen that the  $\vec{E}$  wave that travels in a conducting medium, its amplitude is attenuated by the factor  $e^{-az}$ , as shown in Fig. 2.1.
- Similarly, the Helmholtz or wave equation for  $\vec{H}$  (2.17) can be solved to have:

$$H_{ys}(z) = H^{+}e^{-\gamma z} + H^{-}e^{\gamma z} = H^{+}e^{-\alpha z}e^{-j\beta z} + H^{-}e^{\alpha z}e^{j\beta z}$$

$$E_{x}(z,t)$$

$$E^{+}e^{-\alpha z}\cos(\omega t - \beta z)$$

$$\lambda$$
(2.24)

Fig. 2.1 The electric field wave travels in a conducting medium.

#### 2.3.3 Basic plane wave parameters

#### 1. Phase velocity, v<sub>p</sub>

Consider the positive traveling wave described by Eq. (2.22) as:

$$E_{x}(z,t) = E^{+}e^{-\alpha z}\cos(\omega t - \beta z)$$

To maintain a fixed point on the wave, the phase is constant, therefore:

phase = 
$$(\omega t - \beta z)$$
 = constant (2.25)

• The velocity of the wave is called the phase velocity, because it is the velocity at which a fixed phase point on the wave travel and is given by dz/dt:

$$\frac{d}{dt}(\omega t - \beta z) = 0 \to \omega - \beta \frac{dz}{dt} = 0$$

So, the phase velocity, v<sub>p</sub> is given by:[rad/m]

 $\omega$  in [rad/s] and  $\beta$  in [rad/m]

$$v_p = \frac{dz}{dt} = \frac{\omega}{\beta} \ [m/s] \tag{2.26}$$

#### 2.3.3 Basic plane wave parameters (continued)

#### 2. The Wavelength, $\lambda$

 $\beta$  in [rad/m]

• The wavelength  $\lambda$ , is defined as the distance between two successive maximum (or minima, or any other reference points) on the wave, at a fixed instant of time, thus:  $[(\omega t - \beta z) - (\omega t - \beta (z + \lambda))] = 2 \pi \rightarrow \beta \lambda = 2\pi$ 

$$\lambda = \frac{2\pi}{\beta} \quad [m] \quad and \quad v_p = \frac{\omega}{\beta} \quad \Rightarrow \lambda = \frac{v_p}{f} \quad [m]$$
 (2.27)

#### 3. The Wave Impedance or medium intrinsic impedance, η

- In section 2.3.2, one of wave equations is solved [(2.16) for electric field E or (2.17) for the magnetic field H.
- In general, whenever E or H is known, the other filed vector can be readily found by using one of Maxwell's curl equations.
- Thus applying (2.10) to the electric field obtained (2.18) gives:  $\nabla x \vec{E}_s = -j\omega\mu\vec{H}_s$

$$Curl\vec{E}_{s} = \begin{vmatrix} \hat{a}_{x} & \hat{a}_{y} & \hat{a}_{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_{xs} & E_{ys} & E_{zs} \end{vmatrix} = \begin{vmatrix} \hat{a}_{x} & \hat{a}_{y} & \hat{a}_{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E_{xs}(z) & 0 & 0 \end{vmatrix} = 0\hat{a}_{x} - \frac{\partial E_{xs}(z)}{\partial z}\hat{a}_{y} + 0\hat{a}_{z} = -j\omega\mu\vec{H}_{s}$$

#### 2.3.3 Basic plane wave parameters (continued)

#### 3. The Wave Impedance or medium intrinsic impedance, η

- Which is greatly simplified for a single  $E_x$  component varying only with z and  $H_x = H_z = 0$ , we get:  $\frac{dE_{xs}(z)}{dz} = -j\omega\mu H_{ys}$
- Substitute using Eq.(2.21) for E<sub>sx</sub>, we have:  $E_{xs}(z) = E^+ e^{-\gamma z} = E^+ e^{-\alpha z} e^{-j\beta z}$

$$H_{ys} = -\frac{1}{j\omega\mu} (-\gamma) E^{+} e^{-\gamma z} = \frac{\gamma}{j\omega\mu} E^{+} e^{-\alpha z} e^{-j\beta z}$$
 (2.28a)

In real instantaneous form, (2.28a) becomes:

$$H_{y}(z,t) = \frac{\gamma}{j\omega\mu} E^{+} e^{-\alpha z} \cos(\omega t - \beta z) \implies E_{x}(z,t) = \eta H_{y}(z,t)$$
 (2.28b)

Generally, The magnetic field intensity, H is related to electric field intensity, E by:

$$\vec{E} = \eta \vec{H}$$
  $\Rightarrow \vec{H} = \frac{\vec{E}}{\eta}$   $\eta = \frac{|\vec{E}|}{|\vec{H}|} = \frac{in [V/m]}{in [A/m]} \Rightarrow in [\Omega]$  (2.28c)

• Where η is called wave impedance for the plane wave or the intrinsic impedance for the general medium is given by:

$$\eta = \frac{j\omega\mu}{\gamma} \ [\Omega] \tag{2.29}$$

#### 2.3.3 Basic plane wave parameters (continued)

#### 4. Skin Depth

 $\alpha$  in [Np/m]

• From Eq. (2.22), it is seen that as  $\vec{E}$  and  $\vec{H}$  waves travel in a conducting medium, its amplitude is attenuated by the factor  $e^{-\alpha z}$ . The distance  $\delta$ , through which the wave amplitude decreases by a factor  $e^{-1}$  (about 37%) is called skin depth or penetration depth of the medium as shown in Fig. 2.3.

$$E^+e^{-\alpha\delta} = E^+e^{-1} = 0.368E^+$$

$$\alpha \delta_s = 1 \implies \delta_s = \frac{1}{\alpha} [m] (2.30)$$

The skin depth is a measure of the depth to which an electromagnetic wave can penetrate the medium.

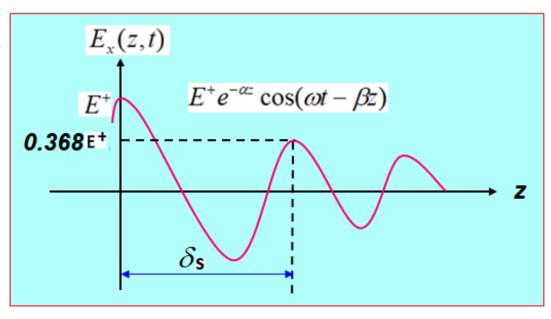


Fig. 2.3 Skin depth

**Example 2.1:** A uniform plane wave is propagating + z direction in in a medium  $(\varepsilon_r = 72, \, \mu_r = 1, \, \sigma = 4 \, (\text{S/m}))$ , If the operating frequency, f = 1 GHz and its electric field intensity at z = 0 and t = 0 is  $|\vec{E}|$  = 10 mV/m and directed toward the x- axis. Determine and calculate the following:

- a) The attenuation constant,  $\alpha$  and phase constant,  $\beta$  also the wave parameters: phase velocity,  $v_p$ , wavelength,  $\lambda$ , intrinsic impedance,  $\eta$  and skin depth  $\delta_s$ .
- b) Write the expressions for the electric and magnetic field intensities  $ec{E}$  and  $ec{H}$  .

#### **Solution**

a) The angular frequency,  $\omega = 2\pi f = 2\pi \times 10^9 = 2\pi G$  rad/s.

$$\frac{\sigma}{\omega\varepsilon} = \frac{4}{2\pi x 10^9 \, x72(1/36\pi) x 10^{-9}} = \frac{4x36\pi}{2\pi x 72} = 1$$

The complex propagation constant,  $\gamma$  for this medium is given by Eq. (2.16):

$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{\sigma}{\omega\varepsilon}} = j2\pi x 10^9 \sqrt{1x4\pi x 10^{-7} x 72x(1/36\pi)x 10^{-9}} \sqrt{1 - j1}$$

$$= j2\pi x 10^9 x 2.83x 10^{-8} x \sqrt{\sqrt{2}e^{-j\pi/4}} = j1066.88[1.18\cos(-22.5^\circ) - j1.18\sin(22.5^\circ)]$$

$$= 481.77 + j1163.09$$
so
$$\alpha = 481.77 \ Np/m \ and \ \beta = 1163.09$$

**Example 2.1 Solution:** The wave parameters are calculated as follows:

From Eq. (2.26), the phase velocity,  $v_p$  is given by:

$$v_p = \frac{\omega}{\beta} = \frac{2\pi x 10^9}{1163.09} = 0.054x 10^8 [m/s]$$

From Eq. (2.27), the wavelength,  $\lambda$  is given by:  $\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{1163.09} = 5.4 \text{ [mm]}$ 

$$\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{1163.09} = 5.4 \ [mm]$$

From Eq. (2.29), the intrinsic impedance,  $\eta$  is given by:

$$\eta = \frac{j\omega\mu}{\gamma} = \frac{j2\pi x 10^9 x 4\pi x 10^{-7}}{481.77 + j1163.09} = \frac{j2\pi x 10^9 x 4\pi x 10^{-7}}{481.77 + j1163.09}$$

$$= \frac{j7898.68}{481.77 + j1163.09} = 5.78 + j2.38 = 6.25 \angle 22.4^{\circ} [\Omega]$$

From Eq. (2.30), the skin depth  $\delta_s$  is given by:  $\delta_s = \frac{1}{\alpha} = \frac{1}{481.77} = 2.08 \ [mm]$ 

b) From Eq. (2.22), The electric field intensity  $\vec{E}$  is given by:

$$\vec{E}(z,t) = E^+ e^{-\alpha z} \cos(\omega t - \beta z) \hat{a}_x = 10 \ e^{-481.77z} \cos(\omega t - 1163.0z) \hat{a}_x \ mV / m$$

From Eq. (2.28), The magnetic field intensity  $\vec{H}$  is given by: :

$$\vec{H}(z,t) = \frac{E_x(z,t)}{\eta} \hat{a}_y = \frac{10 \ e^{-481.77z} \cos(\omega t - 1163.0z)}{6.25 \angle 22.4^o} \hat{a}_y \qquad \theta^{rad} = \frac{\theta^o}{180} \pi$$

$$= 1.6e^{-481.77z} \cos(\omega t - 1163.0z - 0.124\pi) \hat{a}_y \ mA/m$$

# 2.4 Solution of Wave Equations in Free Space

• When considering electromagnetic waves, in free space, it is called sourceless  $(\rho_v = 0, J = 0, \sigma = 0, \mu = \mu_o \text{ and } \epsilon = \epsilon_o)$ . Under these conditions, the complex propagation constant,  $\gamma$  for the medium defined Eq. (2.16c) is given by:

$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{\sigma}{\omega\varepsilon}} = j\omega\sqrt{\mu_o\varepsilon_o}\sqrt{1 - j\frac{0}{\omega\varepsilon}} = j\omega\sqrt{\mu_o\varepsilon_o}$$
 (2.30a)

- Therefore,  $\alpha = 0$ ,  $\beta = \omega \sqrt{\mu_o \varepsilon_o} = k_o$  and  $\gamma^2 = -\omega^2 \mu_o \varepsilon_o$  (2.30b)
- The waves equations (2.16) and (2.17) become:  $\alpha = 0$  ,  $\beta = \omega \sqrt{\mu_0 \varepsilon_0} = \omega/c$

$$\nabla^2 \vec{E}_s - \gamma^2 \vec{E}_s = 0 \quad \Rightarrow \nabla^2 \vec{E}_s + \omega^2 \mu_o \varepsilon_o \vec{E}_s = 0$$

so 
$$\nabla^2 \vec{E}_s + k_o^2 \vec{E}_s = 0$$
 (2.31) and  $\nabla^2 \vec{H}_s + k_o^2 \vec{H}_s = 0$  (2.32)

The constant  $k_o = \omega \sqrt{\mu_o \varepsilon_o}$  is called *the wave number or propagation* constant.

 The solution to Eq.(2.31) will be forward and backward propagating waves having the phasor solutions in the form:

$$E_{xs}(z) = E^{+}e^{-\gamma z} + E^{-}e^{\gamma z} = E^{+}e^{-jk_{o}z} + E^{+}e^{jk_{o}z}$$
 (2.33a)

• The above solution is for time harmonic case at frequency  $\omega$ . In time domain, this result is written as:  $E_{r}(z,t) = \text{Re}[E_{rs}(z)e^{j\omega t}]$ 

$$E_{x}(z,t) = E^{+}\cos(\omega t - k_{o}z) + E^{-}\cos(\omega t + k_{o}z)$$
(2.33b)

• The magnetic field intensity, H is related to electric field intensity, E by:

$$\vec{E} = \eta_o \vec{H} \quad \Rightarrow \vec{H} = \vec{E} / \eta_o \tag{2.34}$$

Note the following characteristics of the forward wave in eq.(2.33):

$$E_{x}(z,t) = E^{+} \cos(\omega t - k_{o} z)$$

- 1. It is time harmonic because we assumed time dependence  $e^{j\omega t}$  to arrive .
- 2. E+ is called the amplitude of the wave and has the same units as E.
- 3. ( $\omega t k_0 z$ ) is the phase (in radians) of the wave; it depends on time t and space variable z.
- 4.  $\omega$  is the angular frequency (rad/s) and  $k_0$  is the phase constant or wave number (rad/m).
- Due to the variation of E with both time t and space variable z, we may plot E
  as a function of t by keeping z constant and vice versa. The plots of
  E(z, t = constant) and E(t, z = constant) are as in Fig. 2.4(a) and (b).
- From Fig. 2.4(a), we observe that the wave takes distance  $\lambda$  to repeat itself and hence  $\lambda$  is called the wavelength (in meters).
- From Fig. 2.4(b), the wave takes time T to repeat itself; consequently T is known as the period (in seconds). Since it takes time T for the wave to travel
- distance  $\lambda$  at the speed v, we expect that:  $\lambda = v_{_p}T = v_{_p} \, / \, f$

 But T = I/f, where f is the frequency of the wave in (Hz).

$$\lambda = v_p T = \frac{v_p}{f} \implies v_p = f\lambda$$

 Because of this fixed relationship between wavelength and frequency, one can identify the position of a radio station within its band by either the frequency or the wavelength. Usually the frequency is preferred.

$$k_o = 2\pi / \lambda_o$$

 Which shows that for every wavelength of distance traveled, a wave undergoes a phase change of 2π radians.

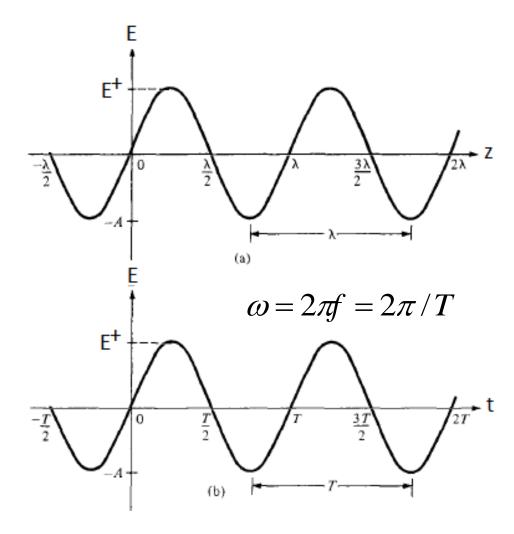


Fig. 2.4 Plot of  $E_x(z,t) = E^+ \sin(\omega t - k_o z)$ : (a) with constant t and (b) with constant z.

#### Basic plane wave parameters

#### 1. Phase velocity, v<sub>p</sub>

• The phase velocity,  $v_p$  in the general medium is given by Eq. (2.26) as:

$$v_p = \frac{dz}{dt} = \frac{\omega}{\beta}$$
  $\omega$  in  $[rad/s]$  and  $k_o$  in  $[rad/m]$  (2.26)

since 
$$\beta = k_o = \omega \sqrt{\mu_o \varepsilon_o}$$
  $v_p = \frac{\omega}{k_o} = \frac{\omega}{\omega \sqrt{\mu_o \varepsilon_o}} = \frac{1}{\sqrt{\mu_o \varepsilon_o}} [m/s]$  (2.35)

For free space, the phase velocity is given by:

$$v_p = \frac{1}{\sqrt{\mu_o \varepsilon_o}} = \frac{1}{\sqrt{4\pi x 10^{-7} x \frac{1}{36\pi} x 10^{-9}}} = 3x 10^8 m / \sec = c$$

c is the speed of light in the free space.

#### 2. The Wavelength, $\lambda$

• The wavelength,  $\lambda$  in the general medium is given by Eq. (2.27) as:  $\lambda = \frac{2\pi}{\beta}$  [m] So, for free space,  $\lambda_0$  is given by:

$$\lambda_o = \lambda = \frac{2\pi}{\beta} = \frac{2\pi}{k_o} = \frac{2\pi}{\omega \sqrt{\mu_o \varepsilon_o}} = \frac{2\pi}{2\pi f \sqrt{\mu_o \varepsilon_o}} = \frac{c}{f} \quad [m]$$
 (2.36)

#### Basic plane wave parameters

#### 3. The Wave Impedance or medium intrinsic impedance, n

The intrinsic impedance for the general medium is given by Eq. (2.29):

$$\eta = \frac{j\omega\mu}{\gamma} [\Omega]$$

So, for free space, the intrinsic impedance,  $\eta_0$  is given by:

$$\eta_o = \eta = \frac{j\omega\mu}{\gamma} = \frac{j\omega\mu}{j\beta} = \frac{\omega\mu_o}{k_o} = \frac{\omega\mu_o}{\omega\sqrt{\mu_o\varepsilon_o}} = \sqrt{\frac{\mu_o}{\varepsilon_o}} \left[\Omega\right]$$

$$\eta_o = \sqrt{\frac{\mu_o}{\varepsilon_o}} = \sqrt{\frac{4\pi x 10^{-7}}{10^{-9}/36\pi}} = \sqrt{144\pi^2 x 10^2} = 120\pi \ \Omega = 377 \ \Omega$$

$$(2.37)$$

#### 4. Skin Depth

Skin Depth

The skin depth for the general medium is given by Eq. (2.30):  $\delta_s = \frac{1}{\alpha}$  [m] So, for free space:

$$\delta_s = \frac{1}{\alpha} = \frac{1}{0} = \infty \tag{2.38}$$

Therefore, the electromagnetic wave can propagate without attenuation in the free space.

#### Basic plane wave parameters (continued) Example 2.2

A uniform plane wave in free space its electric field intensity,  $\vec{E}$  is given by:

$$\vec{E} = 0.8\cos(2\pi x 10^8 t - k_o y)\hat{a}_z \ mV/m$$
. Find:

- a) The propagation constant, k<sub>o</sub>.
- b) The wavelength, λ<sub>o</sub>.
- c) The magnetic field intensity,  $|\vec{H}|$  at the point P(0.1, 1.5, 0.4) at t = 8 ns.

**Solution** a) The propagation constant,  $k_0$  is given by:

$$k_o = \omega \sqrt{\mu_o \varepsilon_o} = \frac{\omega}{c} = \frac{2\pi x 10^8}{3x 10^8} = 2.094 \ rad/m$$
 b) The wavelength,  $\lambda_o$  is given by: 
$$\lambda_o = \frac{c}{f} = \frac{3x 10^8}{10^8} = 3 \ m$$

- c) The magnetic field intensity,  $\vec{H}$  is given by (2.28) as:

$$\vec{H} = \frac{0.8x10^{-3}\cos(2\pi x10^8 t - k_o y)}{\eta_o} \hat{a}_x = \frac{0.8x10^{-3}\cos(2\pi x10^8 t - 2.094 y)}{120\pi} \hat{a}_x$$
$$= 2.12\cos(2\pi x10^8 t - 2.094 y) \hat{a}_x \ \mu A/m$$

At the point P(0.1, 1.5, 0.4) at t = 8 ns, we have:

$$|\vec{H}| = 2.12\cos(2\pi x 10^8 x 8x 10^{-9} - 2.094x 1.5)\hat{a}_x = 12.2|\cos 108.03^{\circ}| = 3.68 \ \mu\text{A/m}$$

#### Basic plane wave parameters (continued)

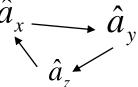
**Example 2.3:** In the free space if the magnetic field intensity,  $\vec{H}$  of a plane wave is given by:  $H(z,t) = 2.65\sin(\omega t - \beta z)\hat{a}_v mA/m$ 

If the working frequency is f = 10 GHz, determine and calculate the following:

- a) The basic plane wave parameters,  $v_p$ ,  $\lambda_o$  and  $\eta_o$ .
- b) The electric field intensity,  $\vec{E}$ .

**Solution** a) The propagation constant,  $k_0$  is given by:

$$v_p = c = \frac{\omega}{\beta} = \frac{2\pi x 10^9}{209.4} = 3x 10^8 \text{ m/sec}$$



The wavelength,  $\lambda_o$  is given by:  $\lambda_o = \frac{2\pi}{\beta} = \frac{2\pi}{k_o} = \frac{2\pi}{209.4} = \frac{c}{f} = 0.03 \ m = 3 \ cm$  The intrinsic impedance,  $\eta_o$  is:

$$\eta = \eta_o = \frac{\omega \mu_o}{\beta} = \frac{2\pi x 10^{10} \, x 4\pi x 10^{-7}}{209.4} = 377 \, \Omega = 120\pi \, \Omega$$

b) The electric field intensity,  $\vec{E}$  is given by (2.28) as:

$$\vec{E}(z,t) = 2.65x\eta_o \sin(\omega t - \beta z)\hat{a}_x = 1.0\sin(2\pi x 10^{10} t - 20.94 z)\hat{a}_x V/m$$

# 2.5 Solution of Wave Equations in a Lossless Medium

• When considering electromagnetic waves, in lossless material where  $\mu$  and  $\epsilon$  real ( $\rho_v = 0$ , J = 0,  $\sigma = 0$  or  $\sigma << \omega \epsilon$ ,  $\mu = \mu_r \mu_o$  and  $\epsilon = \epsilon_r \epsilon_o$ ). Under these conditions, the complex propagation constant,  $\gamma$  for the medium defined Eq.

(2.16c) is given by: 
$$\gamma = \alpha + j\beta = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{\sigma}{\omega\varepsilon}} = j\omega\sqrt{\mu\varepsilon}\sqrt{1 - j\frac{0}{\omega\varepsilon}} = j\omega\sqrt{\mu\varepsilon}$$
 (2.39a)

- Therefore,  $\alpha = 0$ ,  $\beta = \omega \sqrt{\mu \varepsilon} = k$  and  $\gamma^2 = -\omega^2 \mu \varepsilon$  (2.39b)
- The waves equations (2.16) and (2.17) become:

$$\nabla^2 \vec{E}_s - \gamma^2 \vec{E}_s = 0 \quad \Rightarrow \nabla^2 \vec{E}_s + \omega^2 \mu \varepsilon \vec{E}_s = 0$$
 so 
$$\nabla^2 \vec{E}_s + k^2 \vec{E}_s = 0 \quad \text{(2.40)} \quad \text{and} \quad \nabla^2 \vec{H}_s + k^2 \vec{H}_s = 0 \quad \text{(2.41)}$$

The constant  $k = \omega \sqrt{\mu \varepsilon}$  is called **the wave number or propagation** constant.

 The solution to Eq.(2.40) will be forward and backward propagating waves having the phasor solutions in the form:

$$E_{xs}(z) = E^{+}e^{-\gamma z} + E^{-}e^{\gamma z} = E^{+}e^{-jkz} + E^{-}e^{jkz}$$
 (2.42a)

• The above solution is for time harmonic case at frequency  $\omega$ . In time domain, this result is written as:  $E_{r}(z,t) = \text{Re}[E_{rs}(z)e^{j\omega t}]$ 

$$E_{x}(z,t) = E^{+}\cos(\omega t - kz) + E^{-}\cos(\omega t + kz)$$
 (2.42b)

#### 2.5 Solution of Wave Equations in a Lossless Medium (continued)

#### Basic plane wave parameters

The magnetic field intensity, H is related to electric field intensity, E by:

$$\vec{E} = \eta \vec{H} \quad \Rightarrow \vec{H} = \vec{E} / \eta$$
 (2.43)

#### 1. Phase velocity, v<sub>p</sub>

• The phase velocity,  $\vec{v_p}$  in the a lossless medium is given by Eq. (2.26) as:  $v_p = \frac{\omega}{\beta}$  since  $\beta = k = \omega \sqrt{\mu \varepsilon} \ v_p = \frac{\omega}{\omega \sqrt{\mu \varepsilon}} = \frac{1}{\sqrt{\mu \varepsilon}} = \frac{1}{\sqrt{\mu_r \mu_o \varepsilon_r \varepsilon_o}} = \frac{c}{\sqrt{\mu_r \varepsilon_r}} \ [m/s]$  (2.44)

The wavelength,  $\lambda$  in a lossless medium is given by Eq. (2.27) as:  $\lambda = \frac{2\pi}{100}$ 

$$\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{k} = \frac{2\pi}{\omega\sqrt{\mu\varepsilon}} = \frac{2\pi}{2\pi f\sqrt{\mu_r\mu_o\varepsilon_r\varepsilon_o}} = \frac{c}{f\sqrt{\mu_r\varepsilon_r}} = \frac{v_p}{f} \quad [m]$$
 (2.45)

#### 3. The Wave Impedance or medium intrinsic impedance, n

The intrinsic impedance for a lossless medium is given by Eq. (2.29):  $\eta = \frac{J\omega\mu}{\omega}$ 

$$\eta = \frac{\omega\mu}{k} = \sqrt{\frac{\mu}{\varepsilon}} = \sqrt{\frac{\mu_r \mu_o}{\varepsilon_r \varepsilon_o}} = \eta_o \sqrt{\frac{\mu_r}{\varepsilon_r}} = 120\pi \sqrt{\frac{\mu_r}{\varepsilon_r}} \quad [\Omega]$$
4. Skin Depth

Skin Depth

The skin depth for the general medium is given by Eq. (2.30):  $\delta_s = \frac{1}{\alpha} [m]$  $\delta = 1/\alpha = 1/0 = \infty$ 

The electromagnetic wave can propagate without attenuation a lossless medium

#### 2.5 Solution of Wave Equations in a Lossless Medium (continued)

#### Basic plane wave parameters (continued)

**Example 2.4:** A 9.4 GHz uniform plane wave is propagating in polyethylene  $(\varepsilon_r=2.26, \mu_r=1)$  in the + z direction. If the amplitude of the magnetic field intensity,  $\vec{H}$  is 7 mA/m, directed in the y axes and the material is assumed to be lossless, determine and calculate the following:

- a) The basic plane wave parameters,  $v_p$ ,  $\lambda$  and  $\eta$ .
- b) The magnetic and electric field intensities,  $\vec{H}$  and  $\vec{E}$ .

**Solution** a) The propagation constant, k is given by:

$$k = \beta = \omega \sqrt{\mu \varepsilon} = 2\pi x 9.4 \times 10^9 \sqrt{4\pi x 10^{-7} \times 2.26 \times (1/36\pi) 10^{-9}} = 295.965 \text{ rad/m}$$

The phase velocity, 
$$v_p$$
 in the general medium is given by:  

$$v_p = \frac{\omega}{\beta} = \frac{2\pi x 9.4 \times 10^9}{295.965} = 1.9956 \times 10^8 \text{ m/sec}$$

The wavelength,  $\lambda$  is given by:  $\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{k} = \frac{2\pi}{295.965} = 0.021 \ m = 2.1 \ cm$ The intrinsic impedance,  $\eta$  is:

$$\eta = \frac{\omega\mu}{\beta} = \frac{2\pi x 9.4 \times 10^9 \times 4\pi \times 10^{-7}}{295.965} = 251.37 \Omega$$

b) The magnetic and electric field intensities,  $\vec{H}$  and  $\vec{E}$  are given by:

$$\vec{H}(z,t) = 7\cos(\omega t - \beta z)\hat{a}_y = 7\cos(2\pi x 9.4x 10^9 t - 295.965 z)\hat{a}_y mA/m$$

$$\vec{E}(z,t) = 7x\eta\cos(\omega t - \beta z)\hat{a}_x = 1.76\cos(2\pi x 9.4x 10^9 t - 295.695 z)\hat{a}_x V/m$$

#### 2.5 Solution of Wave Equations in a Lossless Medium (continued)

**Example 2.5** A plane wave propagating in lossless dielectric medium ( $\varepsilon_r > 1$ ,  $\mu_r = 1$ and  $\sigma = 0$ ) has an electric field given as  $E_x = E^+ \cos (1.51 \times 10^{10} \, \text{t} - 61.6 \, \text{z})$ . For this wave, determine the:

(a) Wavelength, λ.

- (b) Phase velocity, v<sub>o</sub>.
- (c) Dielectric constant of the medium,  $\varepsilon_r$ . (d) Intrinsic or Wave impedance,  $\eta$

#### **Solution**

(a) By comparison with (2.19) we identify  $\omega = 1.51 \times 10^{10}$  rad /sec and k= 61.6 m<sup>-1</sup>  $\gamma = \alpha + j\beta = j$  61.6 m<sup>-1</sup>, then  $\alpha = 0$  and  $k = \beta = 61.6$  which gives the wavelength  $\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{k} = \frac{2\pi}{61.6} = 0.102m$ λas:

(b) The phase velocity can be found from:

$$v_p = \frac{\omega}{\beta} = \frac{\omega}{k} = \frac{1.51 \times 10^{10}}{61.6} = 2.45 \times 10^8 \, \text{m/sec}$$

(c) This is slower than the speed of light by a factor of 1.225. The dielectric

constant, 
$$\varepsilon_r$$
 of the medium can be found as:  

$$v_p = \frac{c}{\sqrt{\varepsilon_r \mu_r}} = \frac{c}{\sqrt{\varepsilon_r}} \quad gives \quad \epsilon_r = (\frac{c}{\upsilon_p})^2 = (\frac{3.0x10^8}{2.45x10^8})^2 = 1.50$$

(c) The intrinsic or wave impedance, η is given by:

$$\eta = \frac{j\omega\mu}{\gamma} = \frac{\omega\mu}{\beta} = \frac{\omega\mu}{k} = \frac{\eta_o}{\sqrt{\epsilon_r}} = \frac{377}{\sqrt{1.5}} = 307.8 \ \Omega$$

# 2.6 Solution of Wave Equations in a Good Conductor

• When considering electromagnetic waves, in conducting medium ( $\rho_v = 0$ ,  $\sigma \neq 0$ ,  $\mu = \mu_r \mu_o$  and  $\epsilon = \epsilon_r \epsilon_o$  also  $\sigma >> \omega \epsilon$ ). Under these conditions, The waves equations (2.16) and (2.17) are:

$$\nabla^2 \vec{E}_s - \gamma^2 \vec{E}_s = 0 \quad , \quad \nabla^2 \vec{H}_s - \gamma^2 \vec{H}_s = 0 \quad \text{and} \quad \gamma = \alpha + j\beta = j\omega \sqrt{\mu\varepsilon} \sqrt{1 - j\frac{\sigma}{\omega\varepsilon}}$$

 Since σ » ωε metals can be categorized as good conductors, so the one can be neglected with respect to σ/ωε (1 << σ/ωε) and the complex propagation constant γ is given by:

$$\gamma = j\omega\sqrt{\mu\varepsilon}\sqrt{-j\frac{\sigma}{\omega\varepsilon}} = j\sqrt{-j\omega\mu\sigma}$$

but 
$$-j = 1 \angle -90$$
 and  $\sqrt{1 \angle -90^{\circ}} = 1 \angle -45^{\circ}$ 

$$\sqrt{-j} = 1 \angle -45^\circ = (\cos 45^\circ - j \sin 45^\circ) = \frac{1}{\sqrt{2}} (1-j)$$

therefore 
$$\gamma = j\sqrt{-j\omega\mu\sigma} = j\frac{(1-j)}{\sqrt{2}}\sqrt{\omega\mu\sigma} = (1+j)\sqrt{\frac{\omega\mu\sigma}{2}}$$

$$= (1+j)\sqrt{\frac{2\pi f\mu\sigma}{2}} = (1+j)\sqrt{\pi f\mu\sigma}$$
 (2.48a)

Therefore: 
$$\alpha = \beta = \sqrt{\pi f \mu \sigma}$$
  $E_{xs}(z) = E^+ e^{-\gamma z} + E^- e^{\gamma z}$  (2.48b)

• Regardless of the parameter  $\mu$  and  $\sigma$  of the conductor or the frequency of applied field,  $\alpha$  and  $\beta$  are equal. If we again assume only an  $E_x$  component traveling in the +z direction the Eq.(2.22) becomes:  $E_x(z,t) = E^+ e^{-\alpha z} \cos(\omega t - \beta z)$ 

$$E_{x}(z,t) = E^{+}e^{-\alpha z}\cos(\omega t - \beta z) = E^{+}e^{-z\sqrt{\pi f\mu\sigma}}\cos(\omega t - z\sqrt{\pi f\mu\sigma})$$
(2.49)

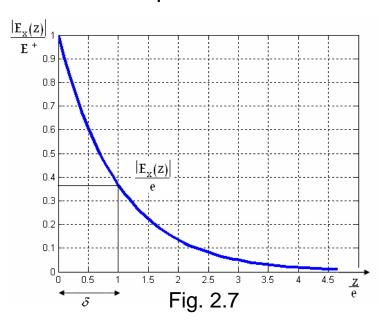
**Definition:** The depth of penetration  $(\delta_s)$  is the distance traveled inside the medium at which the amplitude of the wave attenuates e (e  $\approx$ 2.7) times from its initial level as shown in Fig. 2.7. This is also called skin depth or skin effect.

• From Eq.(2.44) the field amplitude is given by:

$$E_x(z) = E^+ e^{-\alpha z} = E^+ e^{-\sqrt{\pi f \mu \sigma} z} \qquad (2.50)$$
 At  $z = 0$   $|E_x(z)| = E^+$  and at  $z = \delta_s$   $|Ex(z)| = E^+/e$ , therefore: 
$$|E_x(z)| = E^+ e^{-\delta_s \sqrt{\pi f \mu \sigma}} = E^+ e^{-1}$$

• The skin depth or depth of penetration  $\delta_{\text{s}}$  is:

$$\delta_{s} = \frac{1}{\alpha} = \frac{1}{\sqrt{\pi f \mu \sigma}} = \sqrt{\frac{2}{\omega \mu \sigma}}$$



#### Basic plane wave parameters

The magnetic field intensity, H is related to electric field intensity, E by:

$$\vec{E} = \eta \vec{H} \quad \Rightarrow \vec{H} = \vec{E} / \eta$$
 (2.51)

#### 1. Phase velocity, v<sub>p</sub>

• The phase velocity,  $v_p$  in a conductor medium is given by Eq. (2.26) as:  $v_p = \frac{\omega}{\beta}$  since  $\alpha = \beta = \sqrt{\pi f \mu \sigma}$   $v_p = \frac{\omega}{\sqrt{\pi f \mu \sigma}} = \sqrt{\frac{\omega^2}{\pi f \mu \sigma}} = \sqrt{\frac{2\omega}{\mu \sigma}} \ [m/s]$  (2.52)

The wavelength,  $\lambda$  in a lossless medium is given by Eq. (2.27) as:  $\lambda = \frac{2\pi}{\beta}$  $\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{\sqrt{\pi f u \sigma}} [m]$ (2.53)

#### 3. The Wave Impedance or medium intrinsic impedance, n

The intrinsic impedance for a lossless medium is given by Eq. (2.29):  $\eta = \frac{J\omega\mu}{}$ 

$$\eta = \frac{j\omega\mu}{\gamma} = \frac{j\omega\mu}{\sqrt{\pi f\mu\sigma} + j\sqrt{\pi f\mu\sigma}} \quad [\Omega]$$
(2.54)

#### 4. Skin Depth

The skin depth for the general medium is given by Eq. (2.30):  $\delta_s = \frac{1}{\alpha} [m]$   $\delta_s = \frac{1}{\alpha} \left[ \frac{1}{\alpha \pi f \mu \sigma} \right]$  (2 (2.47)

The electromagnetic wave can propagate without attenuation a lossless medium

**Example 2.6** Compute the skin depth ( $\delta_s$ ) at a frequency of 10 GHz of the following materials with  $\mu_r = 1$  and conductivity,  $\sigma$  as indicated:

- a) Aluminum ( $\sigma_{Al} = 3.816 \times 10^7 \text{ S/m}$ ). b) Gold ( $\sigma_{Gl} = 4.098 \times 10^7 \text{ S/m}$ ).
- c) Copper ( $\sigma_{Co} = 5.818 \times 10^7 \text{ S/m}$ ). d) Silver ( $\sigma_{SI} = 6.173 \times 10^7 \text{ S/m}$ ).

**Solution:** Eq. (2.46) gives the skin depth  $\delta_s$  as:

$$\delta_{s} = \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{1}{\pi f \mu_{r} \mu_{o} \sigma}} = \sqrt{\frac{1}{\pi (10^{10})(4\pi x 10^{-7})}} \sqrt{\frac{1}{\sigma}} = 5.03x 10^{-3} \sqrt{\frac{1}{\sigma}}$$

a) For Aluminum: 
$$\delta_{s_{Al}} = 5.03x10^{-3} \sqrt{\frac{1}{3.816x10^7}} = 8.14x10^{-7} m$$

b) For Gold:

$$\delta_{s_{Gl}} = 5.03x10^{-3} \sqrt{\frac{1}{4.098x10^7}} = 7.86x10^{-7} m$$

c) For Copper:

$$\delta_{s_{Co}} = 5.03x10^{-3} \sqrt{\frac{1}{5.813x10^7}} = 6.60x10^{-7} m$$

d) For Silver:

$$\delta_{s_{Sl}} = 5.03x10^{-3} \sqrt{\frac{1}{6.173x10^7}} = 6.40x10^{-7} m$$

**Example 2.7** A plane wave propagating in the sea water medium ( $\varepsilon_r$ =81,  $\mu_r$  = 1 and  $\sigma = 4$  S/m) Consider an incident wave of frequency 1 MHz, find the:

a) Skin depth,  $\delta_s$ .

b) Wavelength, λ.

c) Phase velocity, v<sub>p</sub>.

d) Intrinsic or Wave impedance, η.

#### **Solution**

(a) The loss tangent is given by: 
$$\frac{\sigma}{\omega \varepsilon} = \frac{4}{(2\pi x 10^6)(81)(8.85x 10^{-12})} = 890 >> 1$$

Sea water is therefore a good conductor at 1 MHz, the skin,  $\delta_s$  of Eq. (2.46) is:

$$\delta_{s} = \frac{1}{\alpha} = \frac{1}{\beta} = \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{1}{\pi f\mu_{o}\sigma}} = \sqrt{\frac{1}{\pi(10^{6})(4\pi x 10^{-7})x(4)}} = 0.25 \ m = 25 \ cm$$

(b) the wavelength  $\lambda$  as:

$$\lambda = \frac{2\pi}{\beta} = 2\pi\delta_s = 1.6 \ m$$
 in free space the wavelength  $\lambda_0 = 300 \ m$ 

(c) The phase velocity can be found from (2.35):

$$v_p = \frac{\omega}{\beta} = \omega \delta = 2\pi x 10^6 (0.25) = 1.6x 10^6 \ m/sec$$
 in free space v = c

(d) The intrinsic or wave impedance,  $\eta$  is given by (2.49):  $\gamma = \alpha + j\beta = 4 + j4$ 

$$\eta = \frac{j\omega\mu}{\gamma} = \frac{2\pi x 1x 10^6 x 4\pi x 10^{-7} j}{(4+j4)} = (0.987 + j0.987) \Omega = 1.4 \angle 45^{\circ} \Omega$$

**Example 2.8** A plane wave is given by:  $\vec{E}(z,t) = 0.5 \ e^{-4z} \cos(10^9 t - 4z) \hat{a}_x \ V/m$  Determine the following:

- a) The propagation constant and the wave parameters (V<sub>p</sub>,  $\lambda$ ,  $\eta$  and  $\delta$ <sub>s</sub>)
- b) The magnetic field, associated with the wave in both phasor and time domain representations.

#### **Solution**

a) 
$$\vec{E}(z,t) = 0.5 \ e^{-4z} \cos(10^9 t - 4z) \, \hat{a}_x \, V/m$$
  $\alpha = \beta = 4 \ Then \ \gamma = (4+j4) \ rad/m$ 

$$v_p = \frac{\omega}{\beta} = \frac{10^9}{4} = 2.5x10^8 \ m/sec$$

$$\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{4} = \frac{\pi}{2} \ m = 1.57m$$

$$\eta = \frac{j\omega\mu}{\gamma} = \frac{j10^9 x4\pi x10^{-7}}{(4+j4)} = \frac{10^9 x4\pi x10^{-7}}{(4-j4)} = (50\pi + j50\pi) \, \Omega = 222e^{j\frac{\pi}{4}} \, \Omega$$

$$\delta_s = \frac{1}{\alpha} = \frac{1}{4} = 0.25 \ m$$

$$\vec{E}_s(z) = E^+ e^{-\alpha z} e^{-j\beta z} \, \hat{a}_x \qquad \vec{E}_s(z) = 0.5e^{-4z} e^{-j4z} \, \hat{a}_x \, V/m$$

$$\vec{H}_s(z) = \frac{E^+}{\eta} e^{-\alpha z} e^{-j\beta z} \, \hat{a}_y \qquad \vec{H}_s(z) = \frac{0.5}{222 \ e^{j\frac{\pi}{4}}} e^{-\alpha z} e^{-j\beta z} \, \hat{a}_y = 2.25 \ e^{-4z} \ e^{-j(4z+\frac{\pi}{4})} \, \hat{a}_y \, \text{mA/m}$$

$$\vec{H}(z,t) = 2.25 \ e^{-4z} \cos(10^9 t - 4z - \frac{\pi}{4}) \, \hat{a}_y \, \text{mA/m}$$